

Note that the *instantaneous time-domain* voltage signals  $V_{FG}(t) = V_o^{FG} \cos \omega_o t$  and  $V_{p-mic}(\vec{r}, t) = V_o^{p-mic}(\vec{r}, \omega_o) \cos(\omega_o t + \varphi_p(\vec{r}, \omega_o))$  are *purely real* quantities. We can “*complexify*” these *instantaneous time-domain* quantities by adding *quadrature/imaginary* terms to them:

$$\begin{aligned}\tilde{V}_{FG}(t) &= V_o^{FG} \cos \omega_o t + i V_o^{FG} \sin \omega_o t = V_o^{FG} (\cos \omega_o t + i \sin \omega_o t) = V_o^{FG} e^{i\omega_o t} \quad \text{and:} \\ \tilde{V}_{p-mic}(\vec{r}, t) &= V_o^{p-mic}(\vec{r}, \omega_o) \cos(\omega_o t + \varphi_p(\vec{r}, \omega_o)) + i V_o^{p-mic}(\vec{r}, \omega_o) \sin(\omega_o t + \varphi_p(\vec{r}, \omega_o)) \\ &= V_o^{p-mic}(\vec{r}, \omega_o) \left\{ \cos(\omega_o t + \varphi_p(\vec{r}, \omega_o)) + i \sin(\omega_o t + \varphi_p(\vec{r}, \omega_o)) \right\} = V_o^{p-mic}(\vec{r}, \omega_o) e^{i(\omega_o t + \varphi_p(\vec{r}, \omega_o))}\end{aligned}$$

A {dual-channel} *lock-in amplifier* is *manifestly* a *frequency-domain* device that is routinely used in many types of physics experiments to simultaneously measure the real (*i.e.* in-phase) and imaginary/quadrature (*i.e.* 90° out-of-phase) components of a complex harmonic (*i.e.* single-frequency) signal, *relative* to a *reference* sine-wave signal of the same angular frequency  $\omega_o = 2\pi f_o$ .

In the above example, we could *e.g.* additionally simultaneously connect the microphone’s *time-domain* output signal  $V_{p-mic}(\vec{r}, t) = V_o^{p-mic}(\vec{r}, \omega_o) \cos(\omega_o t + \varphi_p(\vec{r}, \omega_o))$  to the input of the lock-in amplifier and then *also* connect the TTL-level *sync output* of the sine-wave generator to the *reference input* of the lock-in amplifier, which is *phase-locked* to the actual instantaneous {*time-domain*} sine-wave voltage signal  $V_{FG}(t) = V_o^{FG} \cos \omega_o t$  output from the sine-wave generator.

The lock-in amplifier then outputs *frequency-domain* real (“ $X(\omega_o)$ ”) and imaginary (“ $Y(\omega_o)$ ”) components of the complex *p-mic* signal that are respectively in-phase (90° out-of-phase) *relative* to the lock-in amplifier’s *reference input signal* – in this case, the TTL-level *sync signal* output from the sine-wave generator:

$$\begin{aligned}X(\omega_o) &\equiv \text{Re}\{\tilde{V}_{p-mic}(\vec{r}, \omega_o)\} = \text{Re}\left\{V_o^{p-mic}(\vec{r}, \omega_o) e^{i\varphi_p(\vec{r}, \omega_o)}\right\} = V_o^{p-mic}(\vec{r}, \omega_o) \text{Re}\left\{e^{i\varphi_p(\vec{r}, \omega_o)}\right\} \\ &= V_o^{p-mic}(\vec{r}, \omega_o) \text{Re}\left\{\cos \varphi_p(\vec{r}, \omega_o) + i \sin \varphi_p(\vec{r}, \omega_o)\right\} = V_o^{p-mic}(\vec{r}, \omega_o) \cos \varphi_p(\vec{r}, \omega_o) \\ Y(\omega_o) &\equiv \text{Im}\{\tilde{V}_{p-mic}(\vec{r}, \omega_o)\} = \text{Im}\left\{V_o^{p-mic}(\vec{r}, \omega_o) e^{i\varphi_p(\vec{r}, \omega_o)}\right\} = V_o^{p-mic}(\vec{r}, \omega_o) \text{Im}\left\{e^{i\varphi_p(\vec{r}, \omega_o)}\right\} \\ &= V_o^{p-mic}(\vec{r}, \omega_o) \text{Im}\left\{\cos \varphi_p(\vec{r}, \omega_o) + i \sin \varphi_p(\vec{r}, \omega_o)\right\} = V_o^{p-mic}(\vec{r}, \omega_o) \sin \varphi_p(\vec{r}, \omega_o)\end{aligned}$$

Thus, we see that the lock-in amplifier outputs the real (*i.e.* in-phase) and imaginary/quadrature (*i.e.* 90° out-of-phase) components of the *frequency-domain* complex voltage *amplitude* associated with the pressure microphone’s output signal, obtained at the point  $\vec{r}$  in the (complex) sound field of the loudspeaker:

$$\begin{aligned}\tilde{V}_{p-mic}(\vec{r}, \omega_o) &= \text{Re}\{\tilde{V}_{p-mic}(\vec{r}, \omega_o)\} + i \text{Im}\{\tilde{V}_{p-mic}(\vec{r}, \omega_o)\} \\ &= V_o^{p-mic}(\vec{r}, \omega_o) \cos \varphi_p(\vec{r}, \omega_o) + i V_o^{p-mic}(\vec{r}, \omega_o) \sin \varphi_p(\vec{r}, \omega_o) \\ &= V_o^{p-mic}(\vec{r}, \omega_o) \left\{ \cos \varphi_p(\vec{r}, \omega_o) + i \sin \varphi_p(\vec{r}, \omega_o) \right\} = V_o^{p-mic}(\vec{r}, \omega_o) e^{i\varphi_p(\vec{r}, \omega_o)}\end{aligned}$$